## Exercise 1

In continuous conduction mode, obtain the power ratio $P_{o} / P_{T}$ of the buck, boost, buck-boost, Cùk and full-bridge converter. The ratio can be expressed as

$$
\begin{equation*}
\frac{P_{o}}{P_{T}}=\frac{U_{o} I_{o}}{U_{T} I_{T}} \tag{1}
\end{equation*}
$$

where $U_{T}$ is the maximal voltage over the switch, and $I_{T}$ the maximal current in the switch. It is assumed that the current in the choke does not have ripple, i.e. $i_{L}(t)=I_{L}=$ constant

## Solution

In the solution, the ripple in the currents is not taken in account. So, we can use the average currents.

## Buck converter



Figure 1: Buck converter.

The duty ration is given by

$$
\begin{equation*}
D=\frac{t_{o n}}{T s}=\frac{U_{o}}{U_{d}} \tag{2}
\end{equation*}
$$

The current ratio is

$$
\begin{equation*}
\frac{I_{o}}{I_{d}}=\frac{U_{d}}{U_{o}}=\frac{1}{D} \tag{3}
\end{equation*}
$$

The current $I_{T}=I_{o}=I_{L}$ and we get

$$
\begin{equation*}
\frac{P_{o}}{P_{T}}=\frac{U_{o} I_{o}}{U_{d} I_{o}}=D \tag{4}
\end{equation*}
$$

## Boost converter



Figure 2: Boost converter.

The output and input voltage ratio is given by

$$
\begin{equation*}
\frac{U_{o}}{U_{d}}=\frac{1}{1-D} \tag{5}
\end{equation*}
$$

and the current ratio in the

$$
\begin{equation*}
\frac{I_{o}}{I_{d}}=\frac{U_{d}}{U_{o}}=1-D \tag{6}
\end{equation*}
$$

The maximal current in the switch is $I_{T}=I_{d}=I_{L}$ and we get

$$
\begin{equation*}
\frac{P_{o}}{P_{T}}=\frac{U_{o} I_{o}}{U_{d} I_{o}}=\frac{I_{o}}{I_{d}}=1-D \tag{7}
\end{equation*}
$$

## Buck-Boost converter



Figure 3: Buck-Boost converter with one switch.

The ration of the output and input voltage is

$$
\begin{equation*}
\frac{U_{o}}{U_{d}}=\frac{D}{1-D} \tag{8}
\end{equation*}
$$

The current ratio is

$$
\begin{equation*}
\frac{I_{o}}{I_{d}}=\frac{U_{d}}{U_{o}}=\frac{1-D}{D} \tag{9}
\end{equation*}
$$

the maximal current in the switch is $I_{T}=I_{L}=I_{o}+I_{d}$

$$
\begin{equation*}
I_{T}=\left(1+\frac{1-D}{D}\right) I_{d}=\frac{I_{d}}{D} \tag{10}
\end{equation*}
$$

and we get the power ratio

$$
\begin{equation*}
\frac{P_{o}}{P_{T}}=\frac{U_{o} I_{o}}{\left(U_{o}+U_{d}\right) I_{L}}=\frac{1}{1+\frac{U_{d}}{U_{o}} \frac{I_{o}}{D}}=\frac{1}{1+\frac{1-D}{D}} \frac{D(1-D)}{D}=D(1-D) \tag{11}
\end{equation*}
$$

The maximal current of the switch is the same as the current in the inductance. The switch maximal current is not as large as $I_{d}$ because $I_{d}$ is the input average current not its maximal value. The maximal voltage is obtained when the switch is open and at that moment the voltage becomes the sum of the output and input voltage.

## Cùk converter



Figure 4: Cùk converter.
The ratio of the voltage is

$$
\begin{equation*}
\frac{U_{o}}{U_{d}}=\frac{D}{1-D} \tag{12}
\end{equation*}
$$

and the voltage ratio is

$$
\begin{equation*}
\frac{I_{o}}{I_{d}}=\frac{U_{d}}{U_{o}}=\frac{1-D}{D} \tag{13}
\end{equation*}
$$

the maximal current in the switch is $I_{T}=I_{o}+I_{d}$

$$
\begin{equation*}
I_{T}=\left(1+\frac{1-D}{D}\right) I_{d}=\frac{I_{d}}{D} \tag{14}
\end{equation*}
$$

The voltage over the switch is

$$
\begin{equation*}
U_{T}=U_{C 1}=U_{d}+U_{o} \tag{15}
\end{equation*}
$$

and the power ratio is

$$
\begin{equation*}
\frac{P_{o}}{P_{T}}=\frac{U_{o} I_{o}}{\left(U_{o}+U_{d}\right) I_{L}}=\frac{1}{1+\frac{U_{d}}{U_{o}}} \frac{I_{o}}{\frac{I_{d}}{D}}=\frac{1}{1+\frac{1-D}{D}} \frac{D(1-D)}{D}=D(1-D) \tag{16}
\end{equation*}
$$

## Full-bridge converter

The ratio of the voltages is

$$
\begin{equation*}
\frac{U_{o}}{U_{d}}=2 D_{1}-1 \tag{17}
\end{equation*}
$$

and the current ratio is

$$
\begin{equation*}
\frac{I_{o}}{I_{d}}=\frac{U_{d}}{U_{o}}=\frac{1}{2 D_{1}-1} \tag{18}
\end{equation*}
$$



Figure 5: Full-bridge converter.

The thyristor maximal voltage is $U_{d}$ and the converter has four switches. The power ratio

$$
\begin{equation*}
\frac{P_{o}}{P_{T}}=\frac{U_{o} I_{o}}{4 U_{d} I_{o}}=\frac{2 D_{1}-1}{4} \tag{19}
\end{equation*}
$$

In the figure 6 is represented the power ratios.


Figure 6: Power ratio of the converters.

## Exercise 2

Obtain the current $i_{L}$ and the voltage $v_{C}$ in the figure 7. At $t=t_{0}$ the current in the choke is $I_{L 0}$ and the voltage in the capacitor is $V_{C 0}$.


Figure 7: LC circuit and its Laplace equivalent.

The resonance frequency is

$$
\begin{equation*}
\omega_{0}=2 \pi f_{0}=\frac{1}{\sqrt{L C}} \tag{20}
\end{equation*}
$$

and the impedance

$$
\begin{equation*}
Z_{0}=\sqrt{\frac{L}{C}} \tag{21}
\end{equation*}
$$

Following are the useful Laplace-transformations for this exercise.

$$
\begin{align*}
& \frac{s}{s^{2}+a^{2}} \rightarrow \cos (a t)  \tag{22}\\
& \frac{a}{s^{2}+a^{2}} \rightarrow \sin (a t)  \tag{23}\\
& \frac{a^{2}}{s\left(s^{2}+a^{2}\right)} \rightarrow(1-\cos (a t))  \tag{24}\\
& \frac{1}{s} \rightarrow 1 \tag{25}
\end{align*}
$$

## Solution

The current in the capacitor is

$$
\begin{equation*}
i_{c}(t)=C \frac{d V_{c}(t)}{d t} \rightarrow i_{c}(s)=s C v_{c}(s)-C V_{C 0} \tag{26}
\end{equation*}
$$

the voltage over an inductor is

$$
\begin{equation*}
v_{L}(t)=L \frac{d i_{L}(t)}{d t} \rightarrow v_{L}(s)=s L i_{L}(s)-L I_{L 0} \tag{27}
\end{equation*}
$$

Using the figure 7 we get

$$
\begin{gather*}
\frac{V_{d}}{s}-\frac{V_{C 0}}{s}+L I_{L 0}=\left(s L+\frac{1}{s C}\right) i_{L}(s)  \tag{28}\\
i_{L}(s)=\frac{\frac{V_{d}-V_{C 0}}{s}+L I_{L 0}}{s L+\frac{1}{s C}}=\frac{V_{d}-V_{C 0}}{L\left(s^{2}+\frac{1}{L C}\right)}+\frac{s I_{L 0}}{s^{2}+\frac{1}{L C}}=\frac{V_{d}-V_{C 0}}{L\left(s^{2}+\omega_{0}^{2}\right)}+\frac{s I_{L 0}}{s^{2}+\omega_{0}^{2}} \tag{29}
\end{gather*}
$$

and using the Laplace transform inverse we get

$$
\begin{equation*}
i_{L}(t)=I_{L 0} \cos \left(\omega_{0} t\right)+\frac{\left(V_{d}-U_{C 0}\right) \sqrt{L C}}{L} \sin \left(\omega_{0} t\right) \tag{30}
\end{equation*}
$$

The voltage over the capacitor is

$$
\begin{equation*}
u_{c}(s)=\frac{V_{C 0}}{s}+\frac{1}{s C} i_{L}(s)=\frac{V_{C 0}}{s}+\frac{V_{d}-V_{C 0}}{s L C\left(s^{2}+\omega_{0}^{2}\right)}+\frac{I_{L 0}}{C\left(s^{2}+\omega_{0}^{2}\right)} \tag{31}
\end{equation*}
$$

And with the Laplace transform inverse we get

$$
\begin{equation*}
u_{C}(t)=V_{C 0}+\frac{V_{d}-V_{C 0}}{\omega_{0}^{2} L C}\left(1-\cos \left(\omega_{0} t\right)\right)+\frac{\sqrt{L C}}{C} I_{L 0} \sin \left(\omega_{0} t\right) \tag{32}
\end{equation*}
$$

## Exercise 3

A symmetrical square-wave is fed to the RLC of the figure 8 . The frequency of the signal is $f_{s}=f_{0}=50 \mathrm{kHz}$ and the amplitude varies between $\pm V_{d} / 2$ with $V_{d}=20 \mathrm{~V}$. The quality factor of the circuit is $Q=7$ and the components are chosen so that the current component at the fundamental frequency $f_{0}$ has a peak value of 5 A . What is the value of the lowest current harmonic when the resistance $R$ can be about zero at the harmonic frequencies, i.e. when the harmonic number $n>1$. What happens to the current component when the frequency of the input signal rises to 55 kHz ?


Figure 8: RLC circuit.

The quality factor of the circuit is $Q=\frac{\omega_{0} L_{r}}{R}=\frac{1}{\omega_{0} C_{r} R}=\frac{Z_{0}}{R}$.

## Solution

The input signal is on the figure 9 .


Figure 9: Input signal $V$

The Fourier-components of the input signal is

$$
\begin{equation*}
a_{n}=\frac{2 V_{d}}{n \pi} \sin \left(\frac{n \pi}{2}\right) \tag{33}
\end{equation*}
$$

with $a_{0}=0$. The impedance of the circuit is

$$
\begin{equation*}
Z_{n}=R+j\left(n \omega L-\frac{1}{n \omega C}\right) \tag{34}
\end{equation*}
$$

At the resonance frequency the imaginary part of the impedance is zero and we get

$$
\begin{equation*}
\omega=\omega_{0}=\frac{1}{\sqrt{L C}} \tag{35}
\end{equation*}
$$

The components are chosen so that at the resonance frequency $\omega_{0}$, the fundamental peak value is 5 A . Using the equation 33, this current can be written as

$$
\begin{equation*}
\hat{i}=\frac{2 V_{d}}{R \pi} \tag{36}
\end{equation*}
$$

and to obtain a peak value of 5 A , the resistance is $R \approx 2,55 \Omega$. And from the quality factor we get

$$
\begin{equation*}
L=\frac{Q R}{\omega_{0}} \approx 56,74 \mu \mathrm{H} \tag{37}
\end{equation*}
$$

And with the resonance frequency we get,

$$
\begin{equation*}
C=\frac{1}{\omega_{0}^{2} L} \approx 0,178 \mu \mathrm{~F} \tag{38}
\end{equation*}
$$

The impedance $Z_{n}$ can be rewritten as

$$
\begin{equation*}
Z_{n}=R+j \frac{n^{2}\left(\frac{\omega}{\omega_{0}}\right)^{2}-1}{n \omega C} \tag{39}
\end{equation*}
$$

In the exercise, $R \approx 0$ at the harmonic frequencies. if $n>1$

$$
\begin{equation*}
Z_{n} \approx j \frac{n^{2}\left(\frac{\omega}{\omega_{0}}\right)^{2}-1}{n \omega C}=j \frac{n\left(\frac{\omega}{\omega_{0}}\right)^{2}-\frac{1}{n}}{\omega C} \tag{40}
\end{equation*}
$$

When $\omega=\omega_{0}$, the impedance is purely resistive and when $n=3$ the impedance $Z_{3}=47,42 j$ whose module is 18,59 times higher than the resonance impedance $2,55 \Omega$.
The peak value of the current of the 3rd harmonic is

$$
\begin{equation*}
\hat{i}=\frac{\frac{2 V_{d}}{3 \pi}}{\left|Z_{3}\right|} \approx 89 \mathrm{~mA} \tag{41}
\end{equation*}
$$

Despite have a square-wave form signal at the input, the current are quite close to a sinusoidal form. When the circuit quality factor is higher the form of the waveforms becomes more sinusoidal.
When the switching frequency rises to 55 kHz , the impedance at the first harmonic is

$$
\begin{equation*}
Z_{1}=R+j \frac{\left(\frac{\omega}{\omega_{0}}\right)^{2}-1}{\omega C} \approx 2,55+3,43 j \tag{42}
\end{equation*}
$$

The peak value of the fundamental is

$$
\begin{equation*}
\hat{u}_{1}=\frac{2 V_{d}}{\pi} \tag{43}
\end{equation*}
$$

and we get a fundamental peak current of

$$
\begin{equation*}
\hat{i}_{1}=\frac{2 V_{d}}{\pi\left|Z_{1}\right|} \approx 2,98 \mathrm{~A} \tag{44}
\end{equation*}
$$

and the power in the resistor is

$$
\begin{equation*}
P_{55 k H z}=R I^{2}=R\left(\frac{\hat{i}_{1}}{\sqrt{2}}\right)^{2} \approx 11,32 \mathrm{~W} \tag{45}
\end{equation*}
$$

The power in the resistor at the frequency 50 kHz is

$$
\begin{equation*}
P_{50 k H z}=R I^{2}=R\left(\frac{\hat{i}_{1}}{\sqrt{2}}\right)^{2} \approx 31,87=2,82 \times 11,32 \mathrm{~W} \tag{46}
\end{equation*}
$$

